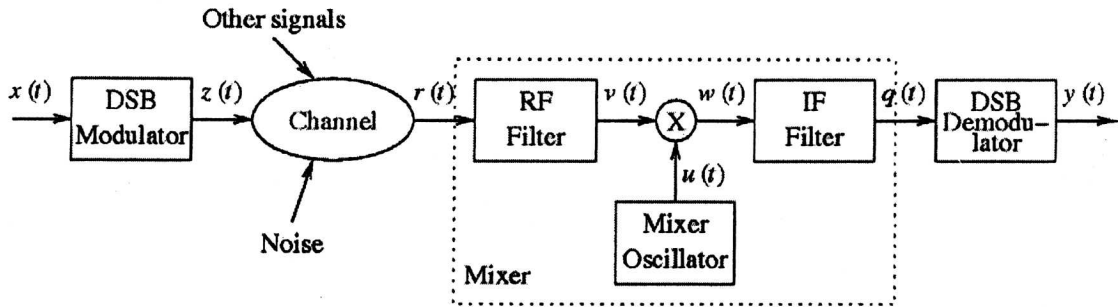


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**Problems:**

1. (Total = 30%) Consider the following transmission system, where DSB stands for “double sideband,” RF stands for “radio frequency” or “radiation frequency,” and IF stands for “intermediate frequency”:



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- (a) (5%) Let the frequency spectrum of the continuous-time signal  $x(t)$  be given by  $X(f) = \frac{1}{2000} \Pi\left(\frac{f}{2000}\right)$  where  $\Pi(f)$  is the “unit pulse function,” i.e.,  $\Pi(f) = 1$  for  $|f| \leq 0.5$  and  $\Pi(f) = 0$  otherwise. Find  $x(t)$  and express it in terms of the sinusoidal function and  $t$ . Note: Do *not* express it *directly* in terms of the sinc function.
- (b) (6%) Following part (a), the DSB modulator output is given by  $z(t) = A_c [1 + ax(t)] \cos(f_c t)$  where  $f_c = 800$  kHz,  $a = 1$ , and  $A_c = 2$ . Sketch the frequency spectrum  $Z(f)$  of  $z(t)$ . Label your plot clearly so that one can reconstruct  $z(t)$  exactly from your plot of  $Z(f)$  with no uncertainty.
- (c) (5%) Following part (b), let the channel and the RF filter be such that  $v(t) = z(t)$ . Let the mixer oscillator output  $u(t) = \cos(f_m t)$  where  $f_m = 900$  kHz. Sketch the frequency spectrum  $W(f)$  of  $w(t)$ . Label it fully as in the case for  $Z(f)$  in part (b).
- (d) (3%) Following part (c), consider the IF filter. Specify a suitable passband for it.
- (e) (5%) Following part (d), let the DSB demodulator be an envelope detector consisting of an ideal diode, a capacitor, and a resistor. Sketch the envelope detector; indicate clearly where its input and output are. *In addition*, give a proper upper bound and a proper lower bound for the  $RC$  constant for the signal considered in this problem.
- (f) (6%) Consider the envelope detector in your answer to part (e). Prove mathematically whether it (from  $q(t)$  to  $y(t)$ ) is a linear system.
2. (Total = 15%) Consider the 16-QAM signaling scheme where each modulated signal can be expressed by

$$s_i(t) = \sqrt{\frac{2}{T_s}} \cdot \left( A_i \cos 2\pi f_c t + B_i \sin 2\pi f_c t \right), \quad \text{for } 0 \leq t < T_s, 1 \leq i \leq 16,$$

where  $A_i$  and  $B_i$  take values on  $\{-3a, -a, +a, +3a\}$  with equal probability,  $T_s$  is the symbol interval, and  $f_c = \frac{1}{T_s}$  is the carrier frequency. The signal is transmitted over an AWGN channel with the signal model  $y(t) = s(t) + n(t)$  where  $s(t) \in \{s_1(t), \dots, s_{16}(t)\}$  and  $n(t)$  is AWGN independent of  $s(t)$  with power spectral density  $N_0/2$ .

- (a) (3%) Find the transmission bit rate of the system and express the average symbol energy in terms of  $a$ .
- (b) (3%) In digital communications, it is common to convert the continuous waveforms to

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equivalent discrete vectors with appropriate choice of basis functions in the vector space in which the inner product of any two waveforms  $f(t)$  and  $g(t)$  is defined as  $\langle f(t), g(t) \rangle \triangleq \int_0^{T_s} f(t)g(t)dt$ .

Suppose we choose  $\phi_1(t) = \sqrt{2/T_s} \cos 2\pi f_c t$  as one basis function in the signal space of 16-QAM. Please find the other basis function  $\phi_2(t)$  that is orthonormal to  $\phi_1(t)$ . Sketch the signal constellation using this basis.

- (c) (4%) Suppose the 4-bit sequence 0000 is assigned to the symbol point  $(-3a, -3a)$  and the 4-bit sequence 0001 is assigned to the symbol point  $(-3a, -a)$ . Please specify the 4-bit sequences to the remaining 14 symbol points using *Gray coding* on the constellation plot in Part (b).
- (d) (5%) The optimum receiver for the communication system with 16-QAM consists of two parallel branches: the in-phase branch  $Y_1 \triangleq \langle y(t), \phi_1(t) \rangle$  and the quadrature branch  $Y_2 \triangleq \langle y(t), \phi_2(t) \rangle$ , where the inner product is defined in Part (b). Please explain why the decoding of the optimum receiver can be carried out by considering these two branches *separately*.

3. (Total = 20%) Alice is sending a binary message  $X$  ( $X=+1$  or  $X=-1$  equally likely) to her new boyfriend Bob through an AWGN communication channel. However, Alice's ex-boyfriend Chuck is so jealous that he is jamming their communications by sending a binary interference  $Z$  ( $Z=+A$  or  $Z=-A$  equally likely) over the same frequency band. The signal  $Y$  that Bob is receiving can be modeled as  $Y = X + Z + N$ , where  $N$  is AWGN with zero mean and variance  $\sigma^2$ , and  $X, Z$  and  $N$  are statistically independent. As Chuck is very upset, he is sending  $Z$  with very large  $A$  which is much greater than 1 and  $\sigma$ . Assume that Bob knows the value of  $A$ , but he doesn't know which one ( $+A$  or  $-A$ ) was sent.

(You may need this: the Q-function is defined as  $Q(x) = \frac{1}{\sqrt{2\pi}} \int_x^\infty e^{-y^2/2} dy$ .)

- (a) (5%) Bob first employs a simple decision strategy by approximating  $Z + N$  as Gaussian with zero mean and variance  $A^2 + \sigma^2$ . What is the optimum decoding strategy, in the sense of minimum probability of decision error, under the Gaussian approximation of  $Z + N$ ? Please find the corresponding probability of error decision. (Express your answer in terms of the Q-function.)
- (b) (5%) Actually the decoding strategy in Part (a) doesn't provide a good decoding quality. So, instead of treating  $Z+N$  as Gaussian, Bob tries the following new decoding strategy:

$$\begin{cases} \hat{X} = +1, & \text{if } Y - A \geq 0 \text{ or } 0 \leq Y + A < A, \\ \hat{X} = -1, & \text{if } Y + A < 0 \text{ or } -A < Y - A < 0, \end{cases}$$

where  $\hat{X}$  denotes the decoded result of  $X$ . Please find the probability of error decision for this new decoding strategy. (Express your answer in terms of the Q-function.)

- (c) (5%) Please interpret the decoding strategy in Part (b). Explain why the decoding strategy in Part (b) performs better than that in Part (a).
- (d) (5%) What is the best decoding strategy, in the sense of minimum probability of error decision, that Bob can come up with? Please provide details of your derivations.

4. (Total = 15%) A communication system is described by a discrete model as  $y_n = x_n * h_n + w_n$  where  $n$  is the time index,  $y_n$  is the received sequence,  $x_n$  is the transmitted symbol

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sequence,  $h_n$  is a length- $L$  sequence representing the channel's impulse response,  $*$  means convolution, and  $w_n$  is AWGN.

- (a) (5%) The purpose of a zero-forcing equalizer is to achieve the zero-ISI condition by operating on the received sequence without considering the AWGN. Please formulate the math equations that may lead to the design of a linear filter as a zero-forcing equalizer.
  - (b) (5%) The purpose of an MMSE equalizer is to minimize the Mean Squared Error (MSE) between the actual transmit symbol and the equalized symbol. Please give a mathematical formulation for the MSE of a linear equalizer.
  - (c) (5%) Assume the channel's impulse response to be  $h_n = [1, -0.5]$ . Find the linear zero-forcing equalizer for it. Please express the equalizer in its impulse response.
5. (Total = 20%) A speech signal is properly sampled with a sampling rate of 8 kHz. Then each sample is quantized with  $q = 2^n$  uniform quantization levels in which  $n$  is the word length.
- (a) (5%) Determine the minimum  $n$  such that the quantization noise is within  $\pm 0.25\%$  of the peak-to-peak full-scale value.
  - (b) (5%) Following part (a), the quantized speech signal (in binary bits) is to be transmitted over a QPSK system. Determine the minimum symbol rate of the QPSK system such that the quantized speech data can be transmitted in real time.
  - (c) (5%) Assume that the quantized speech data first goes through a compression operation such that the data rate is only 20% of the original rate. Does the compressed data stream have lower entropy per data bit than the original data stream has? You need to give detailed explanations.
  - (d) (5%) Furthermore, a rate 2/3 convolutional code is added to enhance the reliability of the transmission. Now determine the minimum symbol rate of the QPSK system to support the real-time transmission of the encoded and compressed speech data.